Linear Signals and Systems — Ph.D. Qualifying Exam Fall 2016

Part 1 (7 pts.)

The frequency response of a *discrete-time* linear time-invariant system is given over the frequency interval $(-\pi, \pi]$ by

$$H(e^{j\omega}) = \begin{cases} 1, & |\omega| \le \pi/3 \\ 0, & \pi/3 < |\omega| \le \pi \end{cases}$$

(1a) (3 pts.) Determine the system's impulse response sequence h[n], expressing your answer in terms of real-valued quantities.

(1b) (4 pts.) If the input to the filter is given at all times $n \in \mathbb{Z}$ by

$$x[n] = \frac{\sin(n\pi/2)}{n\pi} + \sin\left(\frac{2\pi n}{7}\right) + \cos\left(\frac{3\pi n}{7}\right) ,$$

determine the output sequence y[n].

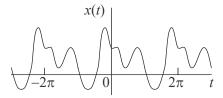
Part 2 (6 pts.)

A continuous-time linear time-invariant system has impulse response

$$h(t) = e^{-rt}u(t)$$

where r > 0. (Here, and in Part 3 below, u(t) is a unit step at t = 0.)

The input to the filter is a periodic signal x(t) having period equal to 2π seconds, as shown below.



If x(t) is given by the (convergent) sum

$$x(t) = \sum_{k=0}^{\infty} A_k \cos(kt + \phi_k) ,$$

obtain a similar equation for the output y(t) using the parameters $\{A_k\}$ and $\{\phi_k\}$.

Part 3 (7 pts.)

A continuous-time, BIBO stable linear time-invariant system has transfer function

$$H(s) = \frac{s}{4-s^2}$$

over an appropriate region of convergence.

(3a) (2 pts.) What is the region of convergence of H(s)? Is the system causal?

(3b) (5 pts.) If the input to the system is x(t) = u(t-1), determine the output y(t).

(1a)

(2*pi)*h[n] is given by the integral of H(exp(j*w))*exp(j*w*n) over w in (-pi,pi]. Thus

$$(2*pi)*h[n] = (exp(j*pi/3) - exp(-j*pi/3))/(j*n)$$

i.e,

h[n] = sin(n*pi/3)/(n*pi)

(1b)

The system is an ideal lowpass filter with cutoff at w = pi/3, gain = 1 and zero delay.

Since 2/7 < 1/3 < 3/7, it follows that the input

$$x'[n] = sin(2*pi*n/7) + cos(3*pi*n/7)$$

produces the output

y'[n] = sin(2*pi*n/7)

The remaining component of x[n], namely

x[n] - x'[n] = sin(n*pi/2)/(n*pi) = (say) g[n]

is the impulse response of a (similar) lowpass filter with cutoff at w = pi/2. Since

$$G(\exp(j^*w))H(\exp(j^*w)) = H(\exp(j^*w)),$$

it follows that

$$y[n] = h[n] + y'[n] = sin(n*pi/3)/(n*pi) + sin(2*pi*n/7)$$

Part 2

 $h(t) = exp(-r^*t)^*u(t)$ has Fourier transform

$$H(j^*W) = 1/(r+j^*W) = (r-j^*W)/(r^2 + W^2)$$

It follows that the input

 $\cos(W^*t + q)$

produces the output

c*cos(W*t + q - arctan(W/r)),

where

 $c = 1/sqrt(r^2 + W^2)$

Since x(t) is the sum of

 $A_k*cos(k*t + q_k)$

over all k, it follows that y(t) is the sum of

$$B_k^{*}\cos(k^{*}t + q_k - \arctan(k/r))$$
,

where

 $B_k = (A_k)/sqrt(r^2 + k^2)$

Part 3

(3a)

 $H(s) = s/(4 - s^2)$ has poles at s = 2, s = -2.

ROC(H) for a BIBO stable system must include s = 0, therefore

 $ROC(H) : -2 < Real\{s\} < 2$

in this case. The system is noncausal (ROC is not right-sided.)

(3b)

If x'(t) = u(t), then X(s) = 1/s with $ROC(X) : 0 < Real\{s\} \le infty$

Then

$$Y'(s) = X'(s)*H(s) = 1/(4 - s^2) = (1/4)/(s+2) - (1/4)/(s-2)$$

with ROC(Y') = ROC(H) (note pole-zero cancelation).

Thus

$$y'(t) = (1/4)*[exp(-2*t)*u(t) + exp(2*t)*u(-t)]$$

$$= \exp(-2^{*}|t|)/4$$
 (OK to modify y'(t) at t = 0)

By time invariance, the response to x(t) = x'(t-1) = u(t-1) is

$$y(t) = \exp(-2^{\ast}|t\text{-}1|)/4$$
 , all t