1. (4 points) The sequence x[n] has a real-valued discrete-time Fourier transform $X(\Omega)$ which is shown below for $-\pi \le \Omega < \pi$.



Find values of Ω_0 in the range $-\pi \leq \Omega_0 < \pi$ for which $y[n] = x[n]e^{j\Omega_0 n}$ is real valued. State the reasons.

2. (4 points) Given the following conditions of a continuous-time, linear and time-invariant system:

(a) The system function is rational, has only two poles at s = -2 and s = 3, and has no more than one zero.

(b) If x(t) = 1, then y(t) = 0

- (i) What is the region of convergence? Justify your answer. (2 points)
- (ii) Is the system stable? Justify your answer. (1 point)
- (iii) Is the system causal? Justify your answer. (1 point)

3. (6 points) A causal discrete-time, linear and time-invariant system has the following system function

$$H(z) = \frac{(z+0.9)(z-0.81)}{(z-j0.9)(z-a)}$$

You are also given that for this filter, when the input is $x[n] = (-1)^n$, the output is $y[n] = 0.1(-1)^n$.

- (i) Determine the location of the pole *a*. (2 points)
- (ii) Does the impulse response of this filter have finite duration? Justify your answer. (1 point)
- (iii) Is the impulse response of this filter real-valued? Justify your answer. (1 point)
- (iv) Would you classify this filter as lowpass, highpass, or bandpass? Justify your answer. (2 points)

4. (6 points) Let x[n] below be the input to an discrete-time, linear and timeinvariant system whose impulse response h[n] is given below. What is the output y[n]?

$$x[n] = \sum_{k=-\infty}^{\infty} \delta[n-2k] \qquad \qquad h[n] = \frac{\sin\left(\frac{\pi n}{2}\right)}{\pi n}$$

Answers:

Part 1. $\Omega_o = \frac{\pi}{4}, \frac{-3\pi}{4}$

Part 2.

- (i) ROC: $-2 < \sigma < 3$, given s=0 is part of the ROC.
- (ii) The system is stable given the ROC includes the $j\omega$ axis.
- (iii) The system is not causal. It is two-sided given the ROC is a vertical strip (not to the right of the rightmost pole and not left of the leftmost pole.).

Part 3.

- (i) a=-j0.9
- (ii) System is IIR since there are poles at frequencies other than 0 and ∞ .
- (iii) It does have a real impulse response since the poles must be complex conjugates of each other in order for H(-1) = 0.1. That is, the length of the vector from $\Omega = \pi$ (*i.e.*, z = -1) to the pole must be $\sqrt{1.81}$ and the angle of the vector must be opposite and equal to the vector from $\Omega = \pi$ to the pole at z = 0.9j.
- (iv) It is a bandpass filter given it has a zero close to $\omega = 0$ and $\omega = \pi$, and it peaks at $\omega = \pi/2$.

Part 4. $X(\Omega) = \pi \sum_{k=-\infty}^{\infty} \delta(\Omega - \pi k)$ and $H(\Omega) = \begin{cases} 1 & |\Omega| < \frac{\pi}{2} \\ 0 & otherwise \end{cases}$, so that $Y(\Omega) = X(\Omega)H(\Omega) = \pi \sum_{k=-\infty}^{\infty} \delta(\Omega - 2\pi k)$ which means that y[n] = 1/2.